## DESCRIPTION

Demonstration circuit2405A shows an LTM ${ }^{\circledR} 2893$ isolating and interfacing an LTC ${ }^{『} 2328$-18. The LTM2893 is a high speed SPI isolator for interfacing read only ADCs with a full complement of control signals. The LTC2328-18 is a low noise, high speed 18-bit successive approximation register (SAR) ADC. Low noise isolated power is delivered to the isolated side with an LT3999 push-pull driver and isolation transformer.

The DC2405A demonstrates the DC and AC operation of the LTC2328-18 without performance degradation with the LTM2893. The Serial Peripheral Interface (SPI) runs
at a maximum 100MHz SCK frequency. The LTM2893 is compatible with many ADCs with SPI clock frequencies up to 100 MHz .
The DC2405A connects to either the DC890 for measurements with PScope ${ }^{\text {TM }}$, or DC590 for measurements with QuikEval ${ }^{T M}$, or DC2026 for measurements with QuikEval and a DC590 sketch or single sample measurements with an example sketch.

Design files for this circuit board are available at http://www.linear.com/demo/DC2405A
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## PGRFORMAOCE SUMMARY Specifications are at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| PARAMETER | CONDITIONS | MIN | TYP | MAX |
| :--- | :--- | :---: | :---: | :---: |
| Input Supply Range | V CC $^{2}$ GND | 4.75 | 5.25 | UNITS |
| Analog Signal Input Range (AIN) |  |  | V |  |
| Clock Frequency (CLK IN) |  | 10 | $\pm 10.24$ | V |
| DC590 Interface Voltage Supply |  | 3.0 | 3.3 | 3.6 |

## DEMO MANUAL DC2405A

## DC2405A COnnection DIAGRAm



Figure 1. DC2405A Setup

## DC2405A JUMPGRS

## DEFINITIONS

JP1 - GND2, Isolated ground connection
JP2 - VCC, 5V Power supply input connection
JP3 - GND, Power supply return connection
JP4 - CFG, Default 0, when CFG = 1 a secondary image in U10 is selected, currently not implemented.

JP5 - OE, Output enable for U10, default ON. When OE is ON, U10 drives the signals to U1 (for use with the DC890, DC590, and DC2026). When OE is OFF the signals to U1 are high impedance from U10 and the interface signals can be driven externally.

JP6 - JTAG, header for factory use only
JP7 - EEPROM, is for factory use only, default WP.

## DEMO MANUAL DC2405A

## DC890 PUICK START PROCEDURE

Check to make sure that all switches and jumpers are set to their default settings as described in the DC2405A Jumpers section and Figure 1 of this manual. The demo board is designed to use the onboard isolated power supply to generate all the required bias voltages. The analog input is DC coupled.

1. Connect the DC2405A to a DC890 USB High Speed Data Collection Board using the edge connector J4.
2. Connect the DC890 to a host PC with a standard USB A/B cable.
3. Apply +5 V and ground to the VCC and GND terminals.
4. Apply a low jitter signal source to J1. Observe the recommended input voltage range for the analog input.
5. Connect a low jitter 100MHz2.5V $\mathrm{V}_{\text {p-psine }}$ wave or square wave to connectorJ2. Note that J2 has a $50 \Omega$ termination resistor to ground. Alternatively, a DC1216A 100MHz fixed frequency clock source board can be used.
6. Run the PScope software (Pscope.exe version K83 or later), which can be downloaded from www.linear.com/ designtools/software. The PScope software should recognize the DC2405A and configure itself automatically.
7. Click the Collect button (see Figure 2) to begin acquiring data. The Collect button then changes to Pause, which can be clicked to stop data acquisition.
Complete PScope software documentation is available from the Help menu. Updates can be downloaded from the Tools menu. Check for updates periodically as new features may be added.


Figure 2. PScope Display Capturing 2kHz Input Tone

## DEMO MANUAL DC2405A

## DC590/DC2026 PUICK START PROCEDURE

IMPORTANT! To avoid damage to the DC2405A, make sure that VCCIO (JP6 of the DC590, JP3 of the DC2026) of the DC590/DC2026 is set to 3.3V before connecting the DC590/DC2026 to the DC2405A.

1. To use the DC590/DC2026 with the DC2405A, it is necessary to apply 5 V and ground to the VCC and GND terminals of the DC2405A.
2. Connect the DC590/DC2026 to a host PC with a standard USB A/B cable.
3. Connect the DC2405A to a DC590/DC2026 USB serial controller using the supplied 14-conductor ribbon cable.
4. Apply a signal source to J 1 . A clock source on J 2 is not necessary.
5. Run the QuikEval software (QuikEval.exe version K107 or later), which is available from www.linear.com/ designtools/software. The correct control panel will be loaded automatically.
6. Click the COLLECT button (Figure 3) to begin reading the ADC.


Figure 3. QuikEval Screenshot Captured with a $50 \Omega$ Terminator on AIN

## DC2405A SETUP

## DC POWER

The DC2405A requires $+5 \mathrm{~V}_{\mathrm{DC}}$ and draws $\sim 265 \mathrm{~mA}$. This current is split between the isolated side and the logic side. The isolated side current consumption is through the DC/ DC power converter supplying the LT1468, input buffer, LTC2328-18 ADC, and the isolated side of the LTM2893. The logic side current supplies the FPGA, clock input path, and the LTM2893.

## CLOCK SOURCE

You must provide a low jitter 2.5V $\mathrm{V}_{\mathrm{p} \text { - }}$ sine or square wave to the clock input J2 for data collection with the DC890. The clock input is AC-coupled so the DC level of the clock signal is not important. A generator, such as the Rohde \& Schwarz SMB100A high speed clock source, is recommended to drive the clock input. Drive J 2 with a 100 MHz clock frequency. The ratio between the clock source and the sampling frequency is 100:1. A 100MHz clock source results in a 1 Msps sampling rate.

## DATA OUTPUT

If not connected to a DC890, parallel data output from this board ( 0 V to 2.5 V by default), can be acquired by a logic analyzer, and subsequently imported into a spreadsheet, or mathematical package depending on what form of digital signal processing is desired. Alternatively, the data can be fed directly into an application circuit. Use pin 50 of

J4, edge connector, to latch the data. The data should be latched using the negative edge of this signal.

## ANALOG INPUTS

The DC2405A analog input AIN is a single-ended input referenced to GND2. AIN has a high input impedance buffer (LT1468) before the LTC2328-18 ADC. The default setup for the DC2405A requires that AIN be driven with a low noise, low distortion generator for SINAD, THD, or SNR testing. Use an analog source such as the Stanford Research DS360 or SR1. Synchronize the clock source to the analog source through an external reference input to generate SNR and SINAD results similar to Figure 2 without windowing.

## LTM2893 DIGITAL INTERFACE

The demo board has an unpopulated header placeholder in-between U10 and U1. All interface signals between the FPGA (U10) and the logic interface of the LTM2893 (U1) are exposed. An external interface may be connected to this header location by setting the OE jumper JP5 off. When JP5 is low, all signals to the LTM2893 will be high impedance from the FPGA. The header placeholder pin pitch is on 0.100 -inch centers.

## DEMO MANUAL DC2405A

## DC2405A DATA COLLECTION

This demo board is tested in-house by attempting to duplicate the FFT plot shown in Figure 2. This involves a 100MHz clock source synchronized with a reference clock to an SR1 sinusoidal generator. The SR1 sinusoidal generator is set at a frequency of 2.01416 kHz . The input signal level is approximately - 1 dBFS . Atypical FFT obtained with DC2405A is shown in Figure 2. Note that to calculate the real SNR, the signal level (F1 amplitude $=-1.006 \mathrm{~dB}$ ) has to be added back to the SNR that PScope displays. With the example shown in Figure 2, this means that the actual SNR would be 94.76 dB instead of the 93.76 dB that PScope displays.
There are a number of scenarios that can produce misleading results when evaluating an ADC. One that is common
is feeding the converter with an input frequency that is a sub-multiple of the sample rate and will only exercise a small subset of the possible output codes. The proper method is to pick an $\mathrm{M} / \mathrm{N}$ frequency for the input sine wave frequency. N is the number of samples in the FFT. M is a prime number between one and $\mathrm{N} / 2$. Multiply $\mathrm{M} / \mathrm{N}$ by the sample rate to obtain the input sine wave frequency. Another scenario that can yield poor results is if you do not have a signal generator capable of ppm frequency accuracy or if it cannot be locked to the clock frequency. You can use an FFT with windowing to reduce the "leakage" or spreading of the fundamental, to get a close approximation of the ADC performance. If an amplifier or clock source with poor phase noise is used, the windowing will not improve the SNR.

## DEMO MANUAL DC2405A

## DC590/DC2026 DATA COLLECTIOn

Due to the relatively low and somewhat unpredictable sample rate of the DC590/DC2026, its usefulness is limited to noise measurement and data collection of slowly moving signals.
To observe measurements from QuikEval, use a DC590 or program a DC2026 with a DC590 sketch from an Arduino IDE. Then launch QuikEval, a typical data capture and histogram are shown in Figure 3.

To observe measurements and see example code for reading and configuring the LTM2893 from the DC2026, run an Arduino IDE and select File > Sketchbook > Part Number $>2000>2800>2893>$ DC2405A. Upload the program to the DC2026 and launch the Serial Monitor from the tools menu. A brief menu will display on the serial monitor output. Selecting 1 and sending to the DC2026 will result in a single conversion and the result will be displayed in the Serial Monitor window.

## DEMO MANUAL DC2405A

## PARTS LIST

| ITEM | QTY | REFERENCE | PART DESCRIPTION | MANUFACTURER/PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | C1 | CAPACITOR, CERAMIC, 3.3nF 5\% 1206 250V NPO | TDK/C3216C0G2E332J085AA |
| 2 | 40 | $\begin{aligned} & \text { C2, C8, C15, C17, C18, C20, } \\ & \text { C23-C28, C34, C45-46, C48-71, C73 } \end{aligned}$ | CAPACITOR, CERAMIC, 100nF 10\% 0402 50V X7R | TDK/C1005X7R1H104K050BB |
| 3 | 9 | C3, C7, C10-13, C43, C47, C72 | CAPACITOR, CERAMIC, 1 1 F 10\% 0603 35V X7R | TDK/C1608X7R1V105K080AC |
| 4 | 4 | C4-5, C29, C36 | CAPACITOR, CERAMIC, 10 $\mu \mathrm{F} 10 \% 120635 \mathrm{~V}$ X7R | TDK/C3216X7R1V106K160AC |
| 5 | 0 | C6, C9 | OPTIONAL |  |
| 6 | 1 | C14 | CAPACITOR, CERAMIC, 47 $\mu \mathrm{F} 20 \% 1210$ 6.3V X7S | TDK/C3225X7S0J476M250AC |
| 7 | 1 | C16 | CAPACITOR, CERAMIC, 2.2 $\mu \mathrm{F}$ 10\% 0603 10V X7R | TDK/C1608X7R1A225K080AC |
| 8 | 4 | C19, C38, C40, C44 | CAPACITOR, CERAMIC, 10 $10 \mathrm{~F} 10 \% 080516 \mathrm{~V}$ X6S | TDK/C2012X6S1C106K085AC |
|  | 2 | C21-22 | CAPACITOR, CERAMIC, 150pF 10\% 1808 250V X7R | MURATA/GA342QR7GF151KW01L |
| 9 | 7 | C30-32, C35, C37, C39, C41 | CAPACITOR, CERAMIC, 10nF 10\% 0402 50V X7R | TDK/C1005X7R1H103K050BB |
| 10 | 1 | C33 | CAPACITOR, CERAMIC, 1 1 F 10\% 1206 100V X7R | TDK/C3216X7R2A105K160AA |
| 11 | 1 | C42 | CAPACITOR, TANTALUM, 47 F 10\% 2413 16V | KEMET/T494C476K016AT |
| 12 | 2 | D1-2 | DIODE, ARRAY, 75V 215mA SOT363 | DIODES INC/BAV99DW-7-F |
| 13 | 2 | J1-2 | CONNECTOR, BNC JACK, EDGE MOUNT | TE CONNECTIVITY/1274727-1 |
| 14 | 1 | J3 | HEADER, $2 \times 72 \mathrm{~mm}$ SHROUDED | MOLEX/87831-1420 |
| 15 | 3 | JP1-3 | HEADER, LOOP $1 \times 2,0.2 \mathrm{~mm}$ | AAVID/125700D00000G |
| 16 | 3 | JP4-5, JP7 | HEADER, $1 \times 3,2 \mathrm{~mm}$ | WURTH/62000311121 |
| 17 | 3 | JP4-5, JP7 | SHUNT, $1 \times 2,2 \mathrm{~mm}$ | WURTH/60800213421 |
| 18 | 1 | JP6 | HEADER, $2 \times 5,0.1 \mathrm{~mm}$ | WURTH/61301021121 |
| 19 | 2 | L1-2 | INDUCTOR, COUPLED, $22 \mu \mathrm{H} 3 \times 3 \mathrm{~mm} 1.9 \Omega$ 0.44A | COILCRAFT/LPD3015-223MRB |
| 20 | 6 | MH1-6 | STANDOFF, NYLON 0.25" | KEYSTONE, 8831 (SNAP ON) |
| 21 | 2 | R3-4 | RESISTOR, $0 \Omega 1 \% 0603$ | VISHAY/CRCW06030000ZOEA |
| 22 | 1 | R2 | OPTIONAL |  |
| 23 | 3 | R9-10, R35 | RESISTOR, 1k $1 \% 0603$ | VISHAY/CRCW06031K00FKEA |
| 24 | 4 | R7-8, R12, R25 | RESISTOR, 33, 1\% 0402 | VISHAY/CRCW040233ROFKED |
| 25 | 1 | R11 | RESISTOR, 49.9 1\% 1206 | VISHAY/CRCW120649R9FKEA |
| 26 | 1 | R6 | RESISTOR, 28k $1 \% 0603$ | VISHAY/CRCW060328KOFKEA |
| 27 | 1 | R5 | RESISTOR, 49.9k $1 \% 0603$ | VISHAY/CRCW060349K9FKEA |
| 28 | 7 | R13, R16, R22, R36-38, R40 | RESISTOR, 4.99k $1 \% 0603$ | VISHAY/CRCW06034K99FKEA |
| 29 | 14 | $\begin{aligned} & \text { R14-15, R17-21, R24, R26, R28, } \\ & \text { R31-33, R34 } \end{aligned}$ | RESISTOR, 10k $1 \% 0402$ | VISHAY/CRCW040210KOFKED |
| 30 | 2 | R1, R23 | RESISTOR, 100k $1 \% 0603$ | VISHAY/CRCW0603100RFKEA |
| 31 | 3 | R27, R29-30 | RESISTOR, 100k $1 \% 0402$ | VISHAY/CRCW0402100KFKED |
| 32 | 1 | R39 | RESISTOR, 2k $1 \% 0603$ | VISHAY/CRCW06032K00FKEA |
| 33 | 11 | RA1-11 | RESISTOR ARRAY, 33k 24 RES 1206 | PANASONIC/EXB-38V330JV |
| 34 | 1 | TR1 | TRANSFORMER, 5kV 2:1 | WURTH/750313626 |
| 35 | 1 | U1 | IC, 100MHz ADC-SPI ISOLATOR | LINEAR TECH/LTM2893CY\#PBF |
| 36 | 1 | U2 | IC, OPAMP | LINEAR TECH/LT1468CS8\#PBF |
| 37 | 1 | U3 | IC, ADC, 1MSPS, 18-BIT, BIPOLAR, $\pm 10.24 \mathrm{~V}$ | LINEAR TECH/LTC2328CMS-18\#PBF |
| 38 | 1 | U4 | IC, PUSH-PULL DC/DC DRIVER, 1A, 1MHz | LINEAR TECH/LT3999EMSE\#PBF |

## PARTS LIST

| ITEM | QTY | REFERENCE | PART DESCRIPTION | MANUFACTURER/PART NUMBER |
| :---: | :---: | :--- | :--- | :--- |
| 39 | 1 | U5 | IC, BUS SWITCH, SPST, SC70-5 | FAIRCHILD SEMI/NC7SZ66P5X |
| 40 | 1 | U6 | IC, INVERTER, SC70-5 | FAIRCHILD SEMI/NC7SZ04P5X |
| 41 | 1 | U7 | IC, D-TYPE POS TRG, MSOP8 | ON SEMI/NL17SZ74USG |
| 42 | 2 | U8-9 | IC, UNBUFFERED INVERTER, SC70-5 | FAIRCHILD SEMI/NC7SVU04P5X |
| 43 | 1 | U10 | IC, FPGA/CPLD 130 I/0 169UBGA | ALTERA/10M08SAU169C8GES |
| 44 | 2 | U11-12 | IC, LD0, 15V, 100mA | LINEAR TECH/LT3060ETS8-15\#PBF |
| 45 | 2 | U13-14 | IC, LDO, 5V, 100mA | LINEAR TECH/LT1761ES5-5\#PBF |
| 46 | 1 | U15 | IC, LDO, 3.3V, 500mA | LINEAR TECH/LT1763CDE-3.3\#PBF |
| 47 | 1 | U16 | IC, BUFFER, TRI-STATE, QUAD, 14TSSOP | NXP/74LVT126PW,118 |
| 48 | 2 | U17-18 | IC, BUS TRANSCEIVER, TRI-STATE, SOT-563 | TI/SN74LVC1T45DRLR |
| 49 | 1 | U19 | IC, EEPROM, 2-KBIT, IC, 8TSSOP | MICROCHIP/24LCO24-I/ST |

## sCHEmATIC DIAGRAM



## SCHEMATIC DIAGRAM



## DEMO MANUAL DC2405A

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